# Information and Coding Theory MA41024/ MA60020/ MA60262 

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Lecture 19
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## Cyclic codes

Recall
$\triangle$ If $\alpha \in F_{q}$ then the minimal polynomial of $\alpha$ over $F_{q}$ is the (lowest degree) irreducible polynomial $f(x) \in F_{p}[x]$ such that $f(\alpha)=0$, where $q=p^{s}$

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$\triangle$ If $\alpha$ has order $e$ (note that in $F_{q}$, the multiplicative group $F_{q} \backslash\{0\}$ is a cyclic group) then the minimal polynomial is

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\prod_{i=0}^{m-1}\left(x-\alpha^{p^{i}}\right)
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where $m$ is the smallest integer such that $p^{m} \equiv 1(\bmod e)$

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$\triangle$ Obviously, a primitive element of $F_{q}$ is a primitive $(q-1)$ th root of unity

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$\rightarrow$ The decoding algo for binary BCH codes was first proposed by Peterson in 1960

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For $n(\geq 3)$ divisor of $q^{m}-1$, for some positive integer, a cyclic code of block length $n$ over the field $F_{q}$, an $(n, k) \mathrm{BCH}$ code with $t$-error-correction for $2 \leq 2 t \leq n-1$ is generated by

$$
g(x)=L C M\left\{m_{m_{0}}(x), m_{m_{0}+1}(x), \ldots, m_{m_{0}+2 t-1}(x)\right\}
$$

where $m_{m_{0}+i}(x), i=0,1, \ldots, 2 t-1$ are minimal polynomials of the $2 t$ successive powers $\alpha^{m_{0}}, \alpha^{m_{0}+1}, \ldots, \alpha^{m_{0}+2 t-1}$ of some $\alpha \in F_{q}$ whose order is $n$ in some extension field $G F\left(q^{m}\right)$.
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Codes with $m_{0}=1$ are called narrow-sense BCH codes

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## Observation

The degree of $g(x) \leq 2 t m$, as there are at most $2 t$ distinct minimal polynomials and each has degree at most $m$

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Set $q=2$.
Suppose $m_{i}(x)$ is the minimal polynomial of $\alpha^{i}$, also let $c(x)=c_{0}+c_{1} x+\ldots+c_{n-1} x^{n-1}$ be a code polynomial with $c_{j} \in F_{2}$

If $\alpha, \alpha^{2}, \ldots, \alpha^{2 t}$ are roots of $c(x)$ then $c(x)$ is divisible by the minimal polynomials $m_{1}(x), m_{2}(x), \ldots, m_{2 t}(x)$ of $\alpha, \alpha^{2}, \ldots, \alpha^{2 t}$, respectively

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Then the generator polynomial of the BCH code is given by

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g(x)=\operatorname{LCM}\left\{m_{1}(x), \ldots, m_{2 t}(x)\right\}
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Nonprimitive BCH codes are defined when $\alpha$ is a nonprimitive element of $G F\left(q^{m}\right)$, and the code length is the order of $\alpha$

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Example $(15,7) \mathrm{BCH}$ code: Let $\alpha$ be a primitive element of $G F\left(2^{4}\right)$ such that $1+\alpha+\alpha^{4}=0$

## Cyclic codes

For any positive integer pair $m, t$ with $m \geq 3, t<n / 2$, there exists a binary BCH code of block length $n=2^{m}-1$, where the number of parity-check bits satisfies $n-k \leq m t$, and the minimum distance $d_{\text {min }} \geq d_{0}=2 t+1$, where $d_{0}$ is called the designed distance of the code.

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5. Find $d_{\min } \geq 2 t+1$ through the parity-check matrix $H$, as discussed for the cyclic code

## BCH code

Minimum distance of BCH code Let $c(x)=c_{0}+c_{1} x+\ldots+c_{n-1} x^{n-1}$ be a code polynomial of a primitive $t$-error correcting BCH code of block length $n=2^{m}-1$

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$\rightarrow$ Suppose $\alpha, \alpha^{2}, \ldots, \alpha^{2 t}$ are roots of $c(x)$, and hence $c(x)$ is divisible by the generator polynomial $g(x)$, the LCM of the $m_{i}(x), 1 \leq i \leq 2 t$

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$$
\mathbf{c}\left[\begin{array}{c}
1 \\
\alpha^{i} \\
\vdots \\
\left(\alpha^{i}\right)^{n-1}
\end{array}\right]=0,1 \leq i \leq 2 t
$$

and hence

$$
\mathbf{c} \cdot H_{i}^{T}=0
$$

where $\mathbf{c}=\left(c_{0}, c_{1}, \ldots, c_{n-1}\right)$ and $H_{i}=\left(1 \alpha^{i} \ldots\left(\alpha^{i}\right)^{n-1}\right), 1 \leq i \leq 2 t$

## BCH code

Now construct the matrix $H$ as follows:

$$
H=\left[\begin{array}{cccc}
1 & \alpha & \ldots & \alpha^{n-1} \\
1 & \alpha^{2} & \ldots & \left(\alpha^{2}\right)^{n-1} \\
\vdots & \vdots & & \vdots \\
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$\rightarrow$ We want to show that any set of $d_{0}-1$ or $2 t$ columns of $H$ cannot be linearly dependent so that the $t$-error-correcting BCH code has minimum distance of at least $d_{0}$ or $2 t+1$

## BCH code

Suppose there exists a codeword whose components consists of the nonzero digits $c_{j_{u}}=1,1 \leq u \leq 2 t$. Then we have

$$
\left(c_{j_{1}}, c_{j_{2}}, \ldots, c_{j_{2 t}}\right) \underbrace{\left[\begin{array}{cccc}
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where $c_{j_{1}}=c_{j_{2}}=\ldots=c_{j_{2 t}}=1$

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However, $|D| \neq 0$, where $|D|$ is called the van der Monde determinant

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## BCH codes

Now evaluating $|D|$ by factoring out $\alpha^{j_{u}}, 1 \leq u \leq 2 t$,

$$
\begin{aligned}
|D| & =\alpha^{j_{1}+j_{2}+\ldots+j_{2 t}}\left|\begin{array}{cccc}
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\end{array}\right| \\
& =\alpha^{j_{1}+j_{2}+\ldots+j_{2 t}} \prod_{v<u}\left(\alpha^{j_{u}}-\alpha^{j_{v}}\right) \neq 0
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Thus any set of $d_{0}-1$ columns is linearly independent and hence the assumption is invalid i.e. the minimum distance of the $t$-error-correcting BCH code is at least the designed distance $d_{0}=2 t-1 \geq d_{\text {min }}$

## BCH code

## Decoding of BCH code computing syndrome

 Suppose that a code polynomial $c(x)$ is transmitted and the received polynomial is $r(x)=c(x)+e(x)$, where $e(x)=e_{0}+e_{1} x+\ldots+e_{n-1} x^{n-1}$ is called the error polynomial.
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Suppose there are $v \leq t$ non-zero coefficients of $e(x)$ in the umknown locations $j_{1}, j_{2}, \ldots, j_{v}$ i.e.

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e(x)=\sum_{j=1}^{v} x^{j_{i}}, 0 \leq j_{i} \leq n-1
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Since $\alpha, \alpha^{2}, \ldots, \alpha^{2 t}$ are roots of each code polynomial, $c\left(\alpha^{i}\right)=0$ for $1 \leq i \leq 2 t$. Thus, from $r(x)=c(x)+e(x)$, we have

$$
r\left(\alpha^{i}\right)=e\left(\alpha^{i}\right), i=1,2, \ldots, 2 t
$$

## BCH codes

Let $s(x)$ denote the syndrome polynomial from the received-word polynomial $r(x)$, given by

$$
\mathbf{S}=\left(S_{1}, S_{2}, \ldots, S_{2 t}\right)=\mathbf{r} \cdot H^{T}
$$

so that

$$
S_{i}=r\left(\alpha^{i}\right)=r_{0}+r_{1} \alpha^{i}+\ldots+r_{n-1} \alpha^{(n-1) i} \in G F\left(2^{m}\right), 1 \leq i \leq 2 t
$$

which corresponds to the syndrome polynomial

$$
s_{i}(x)=s_{0}^{(i)}+s_{1}^{(i)} x+\ldots+s_{n-k-1}^{(i)} x^{n-k-1} \equiv\left(s_{0}^{(1)}, s_{1}^{(i)}, \ldots, s_{n-k-1}^{(i)}\right)
$$

## BCH codes

Let $s(x)$ denote the syndrome polynomial from the received-word polynomial $r(x)$, given by

$$
\mathbf{S}=\left(S_{1}, S_{2}, \ldots, S_{2 t}\right)=\mathbf{r} \cdot H^{T}
$$

so that

$$
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$$

Therefore, each syndrome entry of $\mathbf{S}$ can be computed by dividing $r(x)$ by the minimal polynomial $m_{i}(x)$ for $1 \leq i \leq 2 t$ of $\alpha^{i}$ such that

$$
r(x)=q_{i}(x) m_{i}(x)+p_{i}(x)
$$

## BCH code

Now, the remainder $p_{i}(x)$, where $x=\alpha^{i}$, is the syndrome entry $S_{i}$ since $m_{i}\left(\alpha^{i}\right)=0$. Therefore, computing $r\left(\alpha^{i}\right)$ is equivalent to computinf $p_{i}\left(\alpha^{i}\right)$, and hence

$$
S_{i}=p_{i}\left(\alpha^{i}\right)=r\left(\alpha^{i}\right)=e\left(\alpha^{i}\right), 1 \leq i \leq 2 t
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which further implies that the syndrome vector $\mathbf{S}$ depends only on the error vector $\mathbf{e}$

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Next task Find the error locations

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Next task Find the error locations
Note that

$$
S_{i}=e\left(\alpha^{i}\right)=\sum_{u=1}^{v}\left(\alpha^{j_{u}}\right)^{i}, 1 \leq i \leq 2 t
$$

## BCH codes

Thus we have relations between the syndrome entries and the error parameters $\alpha^{j u}, 1 \leq u \leq v$ :

$$
\begin{aligned}
S_{1}= & \alpha^{j_{1}}+\alpha^{j_{2}}+\ldots+\alpha^{j_{v}} \\
S_{2}= & \left(\alpha^{j_{1}}\right)^{2}+\left(\alpha^{j_{2}}\right)^{2}+\ldots+\left(\alpha^{j_{v}}\right)^{2} \\
\vdots & \vdots \\
S_{2 t}= & \left(\alpha^{j_{1}}\right)^{2 t}+\left(\alpha^{j_{2}}\right)^{2 t}+\ldots+\left(\alpha^{j_{v}}\right)^{2 t}
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When the parameters $\alpha^{j_{u}}, 1 \leq u \leq v$ are determined then the powers $j_{u}$ can finally give the error locations in $e(x)$. These $2 t$ equations are called power-sum symmetric functions

